

# RADIOMETRY

RADIOMETRY - MEASUREMENT OF INCOHERENT ELECTROMAGNETIC ENERGY

- ⊗ ALL MATTER WHETHER GAS, VAPOUR, SOLID OR PLASMA EMITS E.M. ENERGY.
- ⊗ THERMAL EMISSION IS THE DOMINANT PROCESS.
- ⊗ A RADIOMETER IS A VERY SENSITIVE, LOW-NOISE RECEIVER THAT CAN DETECT THIS INCOHERENT E.M. ENERGY.

APPLICATIONS OF RADIOMETRY INCLUDE:

- ⊗ MEASUREMENT OF THE ATMOSPHERE.
- ⊗ SECURITY - OBJECTS HIDDEN BENEATH/ UNDER CLOTHES.
- ⊗ MEDICAL - CLASSIFICATION OF TISSUE TYPES.

THE FUNDAMENTAL QUANTITY IN RADIOMETRY IS THE BRIGHTNESS,  $B$

BRIGHTNESS HAS UNITS OF  $W m^{-2} Hz^{-1} sr^{-1}$

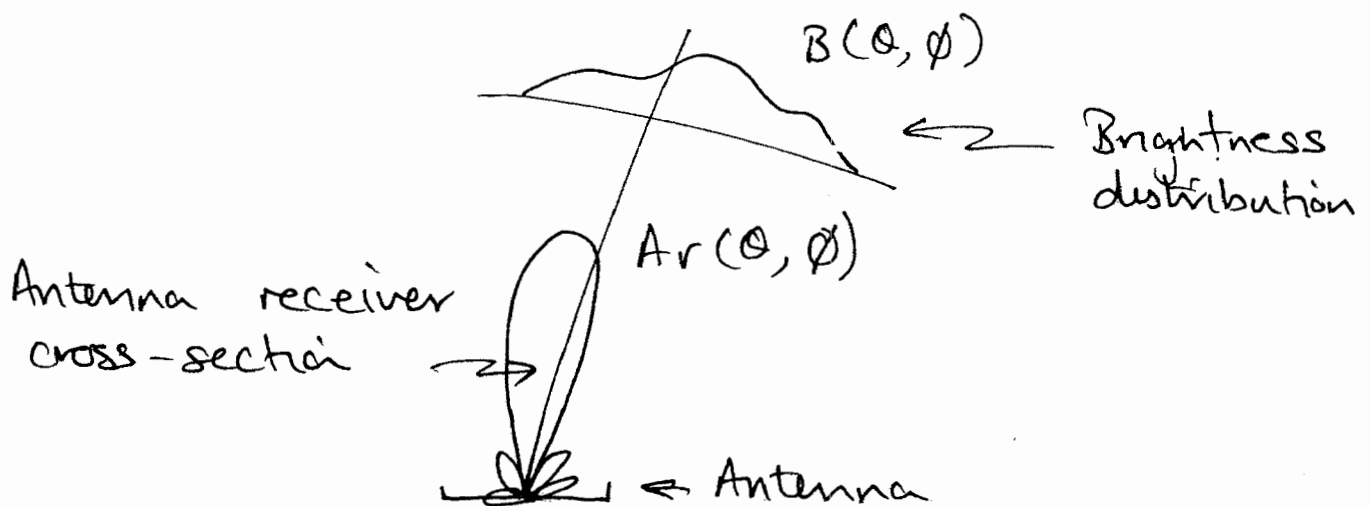
(KNOWN AS RADIANCE IN OPTICAL RADIOMETRY)

- $dP$  - POWER  
 $da$  - AREA  
 $B$  - BRIGHTNESS  
 $d\Omega$  - SOLID ANGLE  
 $df$  - FREQUENCY

$$dP = B \cos\theta da d\Omega df.$$

$B$  - DEPENDS ON DIRECTION -  $B(\theta, \phi)$   
 BRIGHTNESS DISTRIBUTION.

$B$  - DEPENDS ON FREQUENCY -  $B(f)$   
 BRIGHTNESS SPECTRUM.



$$P_r = \frac{1}{2} \int_{4\pi} B(\theta, \phi) A_r(\theta, \phi) d\Omega$$

$1/2$  - ANTENNA IS POLARIZED - INCOHERENT  
 RADIATION IS UNPOLARIZED

IF WE NORMALIZE THE RECEIVING CROSS-SECTION  $A_r$  TO THE MAXIMUM VALUE WE HAVE;

$$P_r = \frac{1}{2} A_{rm} S,$$

$$S = \int_{4\pi} B(\theta, \phi) P_n(\theta, \phi) d\Omega$$

$$A_{rm}(\theta, \phi) = A_{rm} P_n(\theta, \phi)$$

$A_{rm}$  CAN BE WRITTEN IN TERMS OF ANTENNA GAIN ( $G_m$ )

$$A_{rm} = \frac{\lambda^2}{4\pi} G_m.$$

$$S_s = \int_{4\pi} B(\theta, \phi) d\Omega$$

IS THE SOURCE FLUX DENSITY  $W m^{-2} Hz^{-1}$ ,

RADIO ASTRONOMERS CALL THIS BY ANOTHER UNIT THE JANSKY.  $1 \text{ jan} = 1 W Hz^{-1} m^{-2}$   
NAMED AFTER KARL G. JANSKY.

MOST RADIO SOURCES IN RADIO ASTRONOMY ARE OF ORDER  $10^{-26}$  jan.

## BLACKBODY RADIATION

- ⊛ KIRCHHOFF'S LAW: A GOOD ABSORBER OF EM ENERGY IS A GOOD EMITTER.
- ⊛ IN GENERAL RADIATION INCIDENT ON A MATERIAL WILL BE REFLECTED AND ABSORBED.
- ⊛ A BLACKBODY MATERIAL - ABSORBS RADIATION AT ALL FREQUENCIES REFLECTING NONE. - IDEALIZED
- ⊛ A PERFECT ABSORBER IS A PERFECT EMITTER.
- ⊛ ACCORDING TO PLANCK'S RADIATION LAW A BLACKBODY RADIATES UNIFORMLY IN ALL DIRECTIONS WITH A SPECTRAL BRIGHTNESS  $B_f$  GIVEN BY:

$$B_f = \frac{2hf^3}{c^2} \frac{1}{\exp[hf/kT] - 1}$$

- $h$  - PLANCK'S CONSTANT -  $6.63 \times 10^{-34}$  (Js)
- $f$  - FREQUENCY (Hz)
- $k$  - BOLTZMANN'S CONSTANT -  $1.38 \times 10^{-23}$  (Jk<sup>-1</sup>)
- $T$  - TEMPERATURE (K)
- $c$  - VELOCITY OF LIGHT -  $3 \times 10^8$  (ms<sup>-1</sup>)

FOR MICROWAVE & MILLIMETRE WAVES,  
 $hf \ll kT$ .

$$\exp(hf/kT) \approx 1 + hf/kT$$

THIS APPROXIMATION IS CALLED THE  
RAYLEIGH-JEANS LAW;

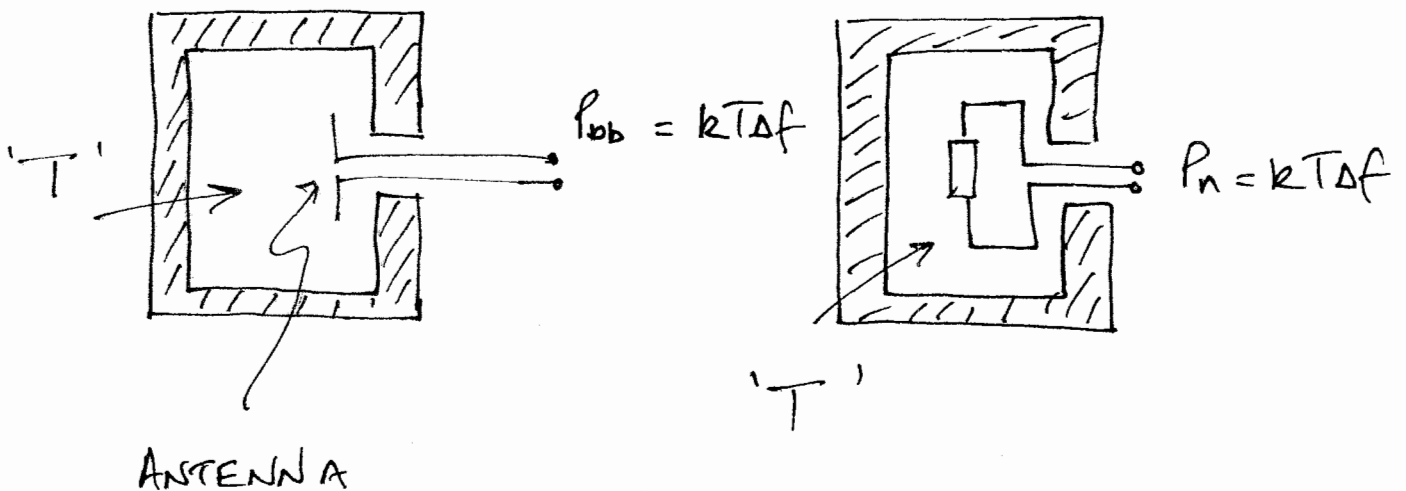
$$B_f = \frac{2k}{\lambda^2} T.$$

STEFAN - BOLTZMANN LAW - TOTAL  
 BRIGHTNESS AT TEMPERATURE T

$$B = \int_0^{\infty} B_f df = \frac{\sigma T^4}{\pi}$$

$\sigma$  - STEFAN BOLTZMANN CONSTANT  
 $5.673 \times 10^{-8} \text{ W m}^{-2} \text{ K}^{-4} \text{ sr}^{-1}$ .

POWER - TEMPERATURE CORRESPONDENCE



$$P_{bb} = kT\Delta f \frac{A_r}{\lambda^2} \int_{4\pi} F_n(\theta, \phi) d\Omega$$

$$R_A = \int_{4\pi} F_n(\theta, \phi) d\Omega = \frac{\lambda^2}{A_r}$$

HENCE  $F_{bb} = kT\Delta f \equiv F_n$ .

### NON BLACKBODY RADIATION

NON BLACKBODIES CALLED "GREY BODIES"

BRIGHTNESS OF BLACKBODY:

$$B_{bb} = B_f \Delta f = \frac{2kT}{\lambda^2} \Delta f.$$

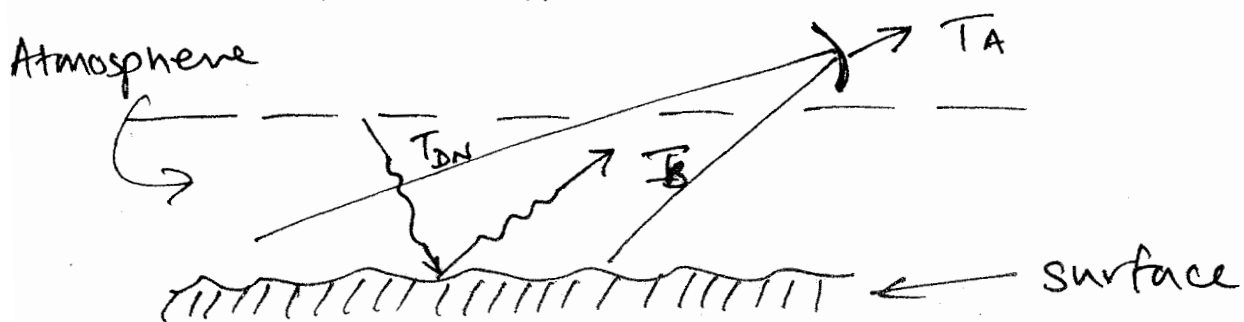
FOR A NARROW BANDWIDTH,  $\Delta f$ .

$$B(\theta, \phi) = \frac{2k}{\lambda^2} T_B(\theta, \phi) \Delta f.$$

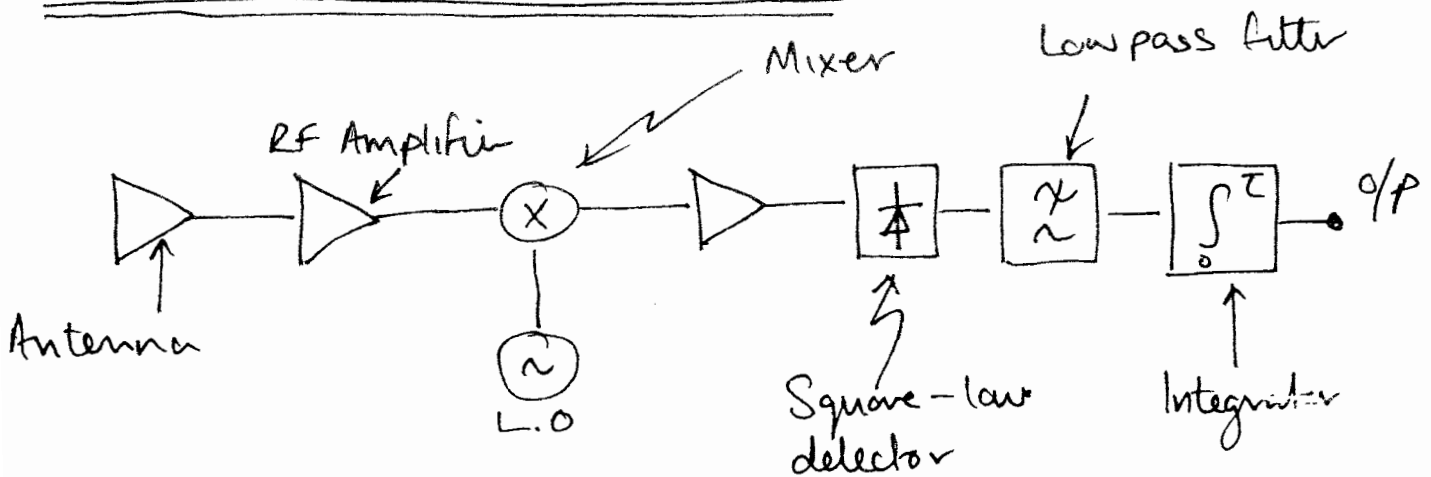
$$e(\theta, \phi) = \frac{B(\theta, \phi)}{B_{bb}} = \frac{T_B(\theta, \phi)}{T}.$$

$e(\theta, \phi)$  - EMISSIVITY  $0 \leq e(\theta, \phi) \leq 1$ .

TO SENSE  $T_B$  WE USE AN ANTENNA AND MEASURE ITS TEMPERATURE,  $T_{AP}$   
IDEALLY  $T_{AP} = T_B$



# TOTAL POWER RADIOMETER



POWER MEASURED CONSISTS OF,  $T_{sys}$ , SYSTEM TEMPERATURE (BACKGROUND) AND SIGNAL TEMPERATURE  $\Delta T$ .

SQUARE-LAW DETECTOR - OUTPUT VOLTAGE PROPORTIONAL TO OUTPUT NOISE POWER

OUTPUT FROM SYSTEM NOISE =  $(k T_{sys} BW)^2$   
 BW - BANDWIDTH OF RECEIVER.  
 AVERAGED BW INTEGRATION =  $BW\tau$ .  
 SIGNAL OUTPUT =  $(k \Delta T BW)^2$

$$V_{sys} \approx \frac{(k T_{sys} BW)^2}{BW\tau} \quad \text{- OUTPUT VOLTAGE SIGNAL.}$$

EFFECTIVE REDUCTION IN NOISE TEMPERATURE FROM  $T_{sys}$  TO  $T_{sys} / (BW\tau)^{1/2}$

$$\text{SIGNAL OUTPUT } V_s \approx (k \Delta T BW)^2$$

$$V_s = V_{sys} \Rightarrow \Delta T = \Delta T_{min}$$

$$\Delta T_{min} = \frac{T_{sys}}{\sqrt{BW\tau}}$$

IN GENERAL;

$$\overline{V_{out}} = g_{LF} C_d G k T_{sys} BW$$

BW - BANDWIDTH

$T_{sys}$  - TEMPERATURE

$k$  - BOLTZMANN'S CONSTANT

$G$  - SYSTEM GAIN

$C_d$  - DETECTOR SENSITIVITY ( $VW^{-1}$ )

$g_{LF}$  - LOW-PASS FILTER GAIN.

THE PROBLEM:

$$\overline{V_{out}} \propto G T_{sys}.$$

$G_s \rightarrow \Delta G_s + G_s$  AS GAIN VARIES  
LEADS TO INCREASE IN  $T_{sys}$  BY

$$\Delta T_g = \Delta T_{sys} = T_{sys} \left( \frac{\Delta G_s}{G_s} \right)$$

SINCE NOISE FLUCTUATIONS ( $\Delta T_{min}$ ) AND GAIN FLUCTUATIONS ARE UNCORRELATED, TOTAL UNCERTAINTY (RMS.) GIVEN BY;

$$\begin{aligned} \Delta T &= \left[ (\Delta T_{min})^2 + (\Delta T_g)^2 \right]^{1/2} \\ &= T_{sys} \left[ \frac{1}{BW \tau} + \left( \frac{\Delta G_s}{G_s} \right)^2 \right]^{1/2}. \end{aligned}$$

$$\begin{aligned} T_{sys} &= 600K, \quad BW = 100MHz, \quad \tau = 10ms, \quad \Delta G_s / G_s = 0.01 \\ T_A &= 300K \quad \Rightarrow \quad \Delta T_{min} = 0.9K, \quad \Delta T_g = 9K, \quad \Delta T = 9.05K \end{aligned}$$



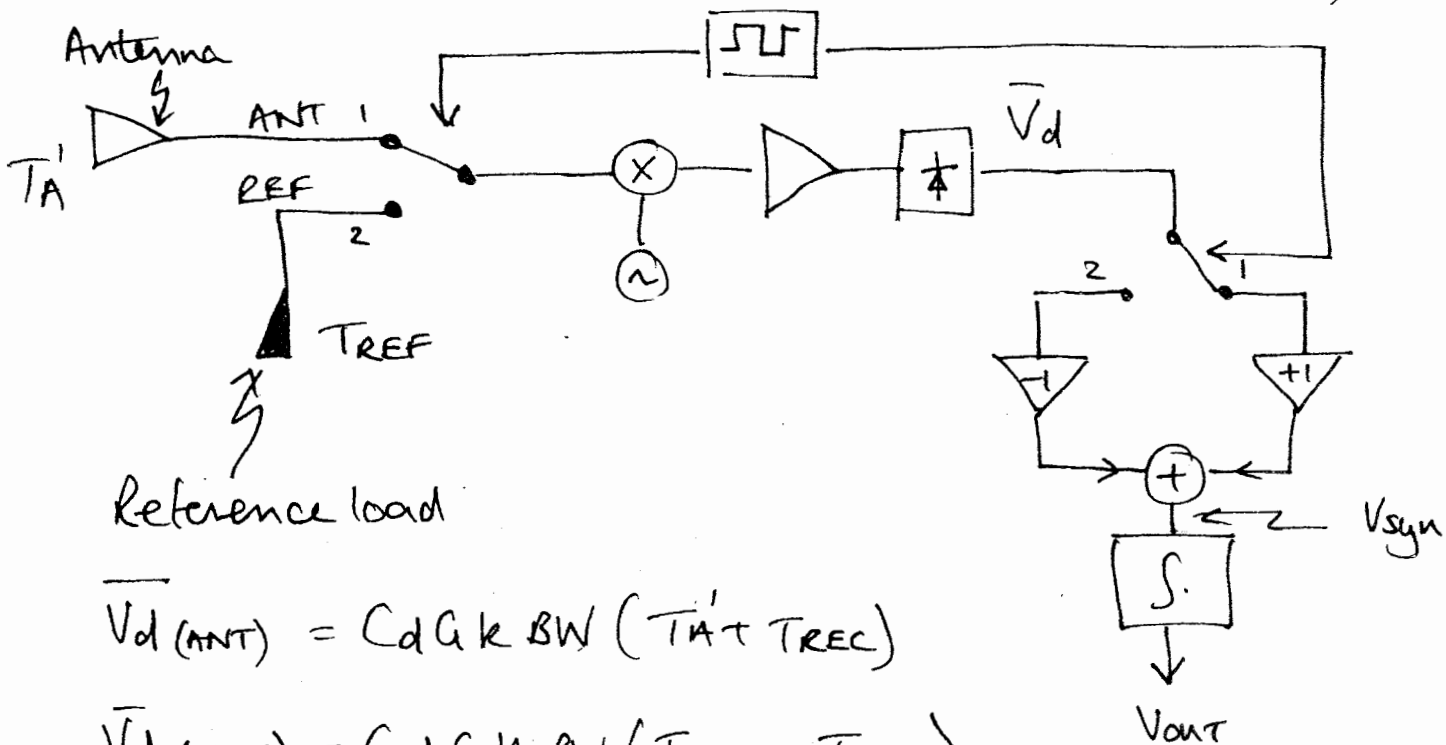
GAIN VARIATIONS CAN DOMINATE SENSITIVITY

→ REDUCE GAIN FLUCTUATIONS!

$\Delta G_s / G_s < 10^{-5}$  CAN BE ACHIEVED, BUT CAN BE COSTLY PARTICULARLY AT MILLIMETER WAVELENGTHS.

## DICKE RADIOMETER

R. H DICKE (1946) SOLVED THIS PROBLEM;



$$\bar{V}_d (ANT) = C_d G K BW (T_A' + T_{REC})$$

$$\bar{V}_d (REF) = C_d G K BW (T_{REF} + T_{REC})$$

$$\begin{aligned} V_{syn} &= \frac{1}{2} (\bar{V}_d (ANT) - \bar{V}_d (REF)) \\ &= \frac{1}{2} C_d G K BW (T_A' - T_{REF}) \end{aligned}$$

IF  $T_A' = T_{REF}$  GAIN FLUCTUATIONS ARE NIL.

$$\Delta T_{MIN} = \left[ \frac{2(T_A' + T_{REC})^2 + 2(T_{REF} + T_{REC})^2}{BW \tau} + \frac{\Delta G_s}{G_s} (T_A - T_{REF}) \right]$$

FOR A BALANCED DICKE RADIOMETER

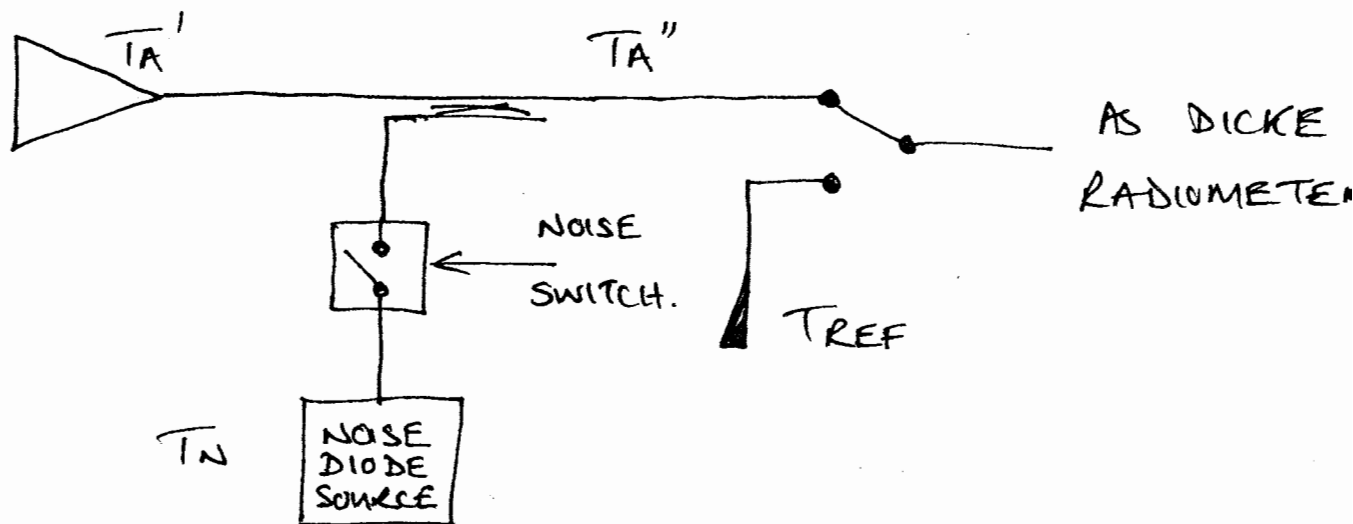
$T_A' = T_{REF}$ , SENSITIVITY CAN BE WRITTEN AS;

$$\Delta T_{MIN} = \frac{2(T_A + T_{REC})}{\sqrt{BWT}}$$

$$T_{SYS} = T_A + T_{REC}$$

$$\Delta T_{MIN} \text{ (DICKE)} = 2 \Delta T_{MIN} \text{ (IDEAL)}$$

PULSED NOISE INJECTION BALANCING



AT BALANCE  $T_A'' = T_{REF} \Rightarrow \overline{V_{syn}} \equiv 0$ .

$$(\overline{T_A' + \alpha T_N}) = T_A'' = T_{REF}$$

ADJUST  $\alpha$  BY PULSING NOISE SWITCH.